AN11062

Broadband DVB-T UHF power amplifier with the BLF888A

Rev. 2 — 1 September 2015

AMPLEON

Application note

Revision history

Rev	Date	Description
AN11062#2	20150901	Modifications • The format of this document has been redesigned to comply with the new identity guidelines of Ampleon. • Legal texts have been adapted to the new company name where appropriate.
AN11062#1	20110530	Initial version

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AN11062#2

Application note **Rev. 2 — 1 September 2015** **2 of 20**

1. Introduction

For the past few years, new product design in the broadcast industry has been dominated by emerging digital modulation standards. A specific example is broadcast television, where many new transmitter systems are currently being set up as part of the conversion to digital terrestrial television.

The BLF888A is a UHF LDMOS power transistor intended for the broadcast transmitter market. The BLF888A can deliver 110 W DVB-T average power over the full UHF band from 470 MHz to 860 MHz at a CCDF of 8 dB (0.01 % probability). Higher average DVB-T power (120 W average) is possible depending on required linearity and predistorter capabilities. The transistor is also capable of handling analog TV (ATV) signals. The average power delivered will be dependent on cooling conditions.

Today, UHF transmitter design focuses on increasing efficiency and output power, while reducing size. To achieve these goals, the next generation of UHF transistors must deliver greater power levels, increased efficiency and higher gain, and the size of application boards needs to be reduced. The BLF888A meets these requirements. The BLF888A has been optimized for extreme low thermal resistance ($R_{th(j-c)}$ = 0.15 K/W) and improved ruggedness (> 40 : 1 at 860 MHz).

This report describes a broadband application incorporating the BLF888A, built on a small form factor board with a total size of 105 mm × 50 mm.

2. Circuit description

The BLF888A broadband application circuit is shown in Figure 1. It is a Class AB common source amplifier. Circuit dimensions are 105 mm × 50 mm (including the transistor). If the connectors are excluded, the dimensions are 95 mm × 50 mm. The PCB material is Taconic RF35 (ε_r = 3.5) with a thickness of 0.76 mm.

AN11062#2

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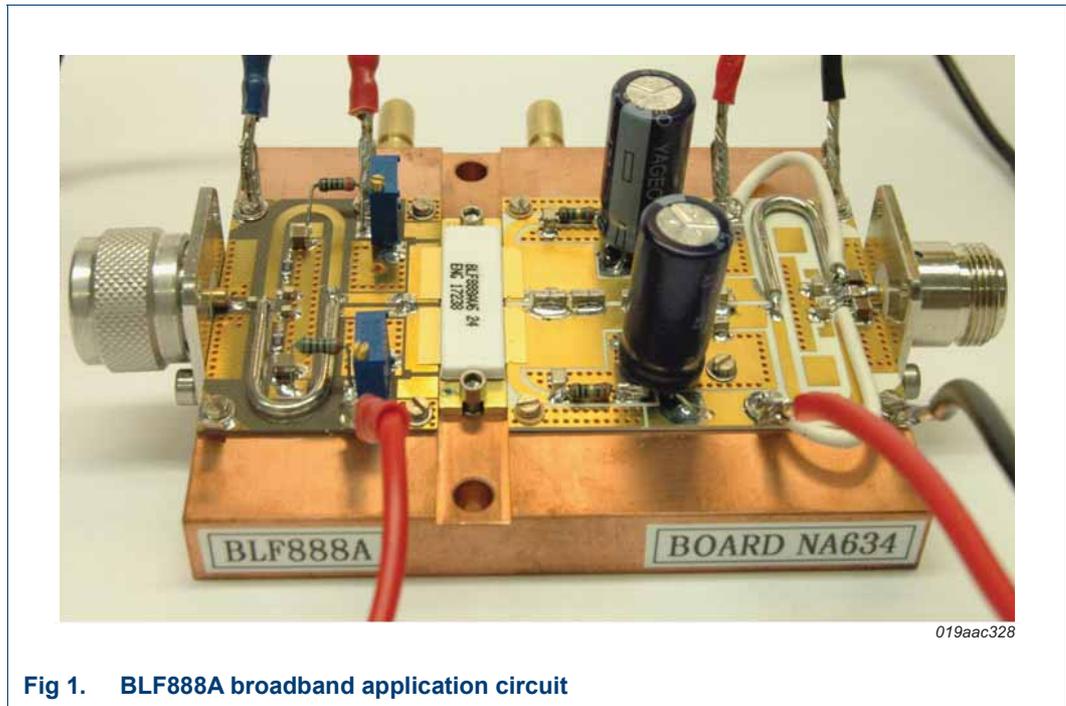**Application note**

Rev. 2 — 1 September 2015

3 of 20

Fig 1.   BLF888A broadband application circuit

A schematic diagram of the application design is shown in Figure 2. The output consists of two microstrip lines per side, L1 and L2, and a balun B1 (25 Ω). C1 to C9 are used to match the transistor impedance to the input impedance of the balun (25 Ω). The balun converts the 25 Ω differential impedance to a 50 Ω asymmetrical impedance at the output. The length of the balun is approximately $\lambda/8$ at the UHF middle frequency. Microstrip line L3 is connected to the differential input of the balun to improve symmetry of the impedance to ground.

L5 in combination with C17 and C18 improves matching at the lower UHF frequencies. C11 and C12 influence the common mode impedance and improve the harmonic behavior. R1 and R2 were added for low frequency damping.

The input also consists of two microstrip lines, L30 and L31, and a balun B2 (25 Ω). C34 and C35 are RF decoupling capacitors for the balun. The transistor gate supply is connected via an RC network consisting of resistors R3 to R6, along with C36 and C37. This RC network performs a damping function that helps improve stability. Microstrip line L32 performs the same function at the input as microstrip line L3 at the output.
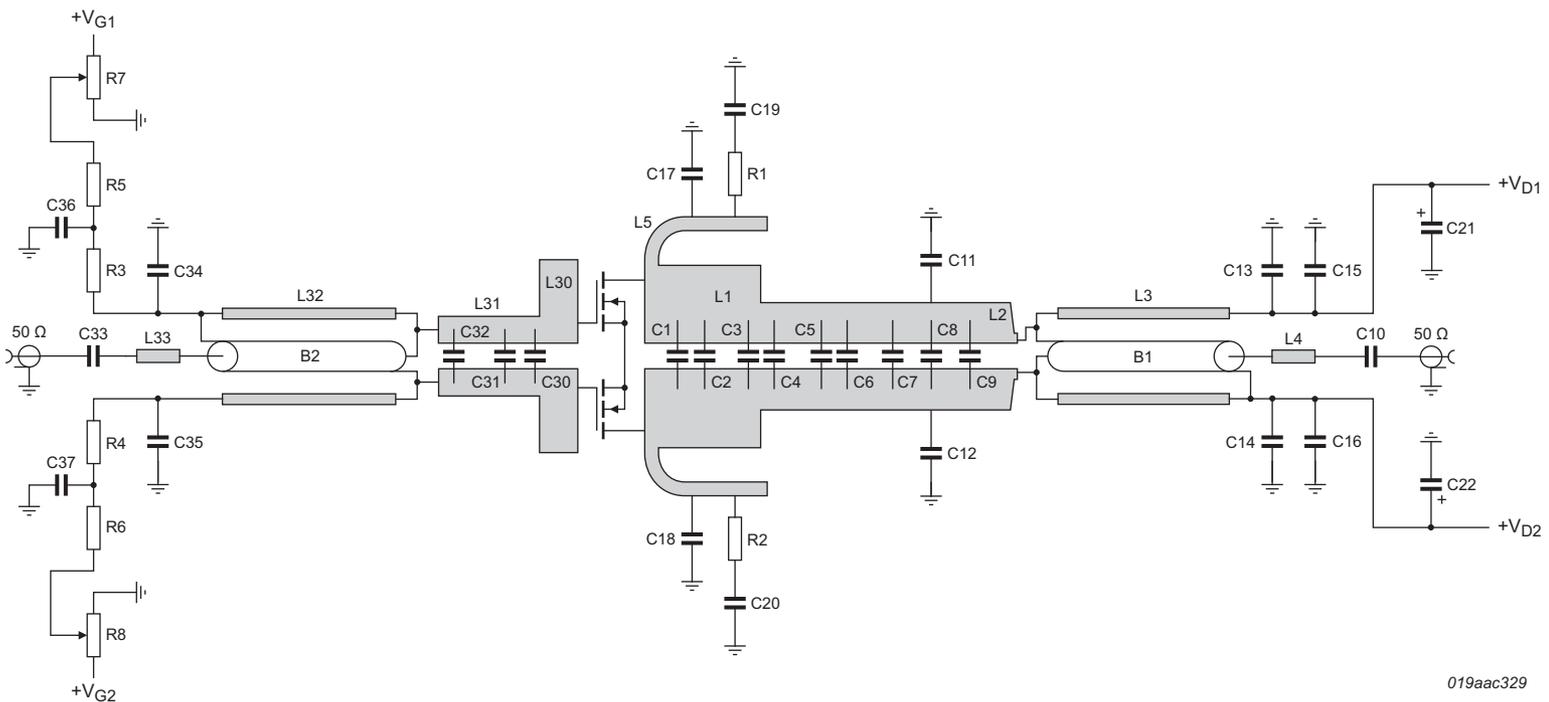
+V$_{G1}$

R7

R5

C36

R3

C34

L32

50 Ω

C33

L33

B2

R4

C35

C37

R6

R8

+V$_{G2}$

L31

C32

C31

L30

C30

L5

C17

C19

R1

C11

L1

C1

C3

C5

C8

L2

C2

C4

C6

C7

C9

C12

C18

R2

C20

C13

C15

+V$_{D1}$

+

C21

L3

L4

C10

50 Ω

C14

C16

C22

+

+V$_{D2}$

B1

019aac329

Fig 2.   Class-AB common-source broadband amplifier

3. Design and Simulation

3.1 BLF888A impedance and simulation data

The impedance data detailed in Figure 3 and Figure 4 was used as a starting point in the development of the broadband circuit.This data is simulated from the equivalent circuit of the transistor. By using the equivalent circuit, impedances can be simulated from the internal drain current source and gate capacitance.

The push-pull and common mode simulation results for the broadband application circuit are shown in Figure 6 and Figure 7 respectively. The (common mode) low impedance in the second harmonic band (940 MHz to 1720 MHz) is of particular significance in the simulation. The resonance peaks in common mode can be shifted via capacitors C11 and C12 (see Figure 2). A high second harmonic impedance will result in significant power and efficiency losses.
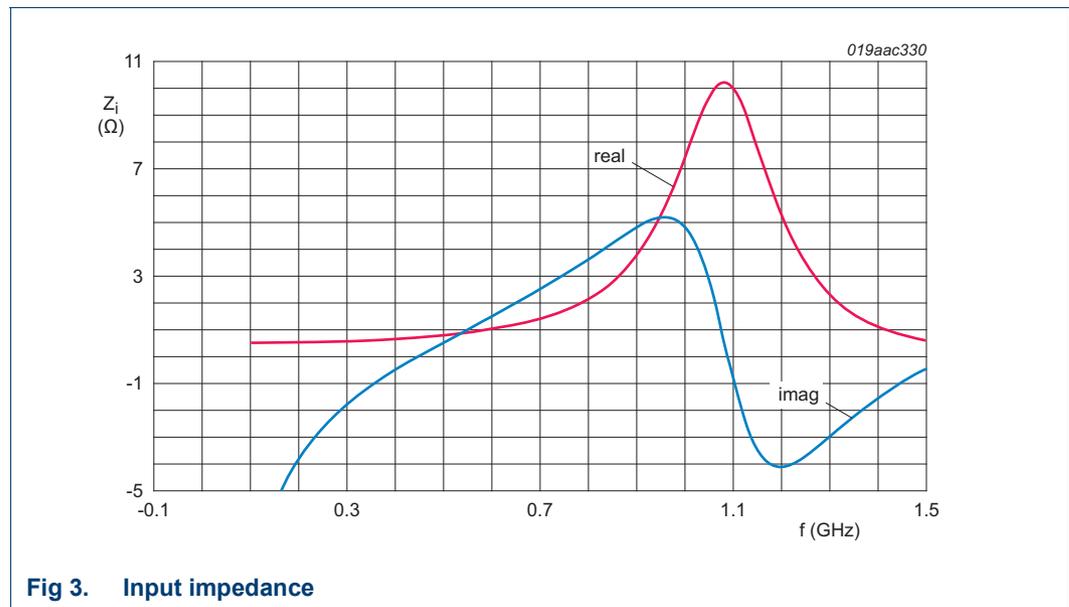
3.1.1 Impedance data
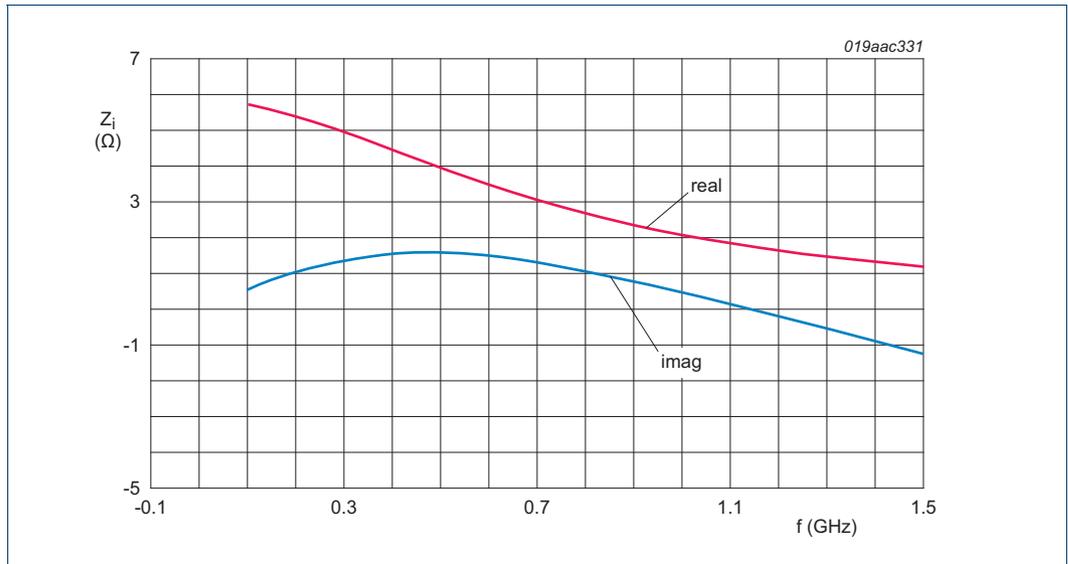

Fig 3.   Input impedance

AN11062#2

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Application note

Rev. 2 — 1 September 2015

6 of 20

Fig 4. **Output impedance; P_O = 110 W DVB-T (total device)**
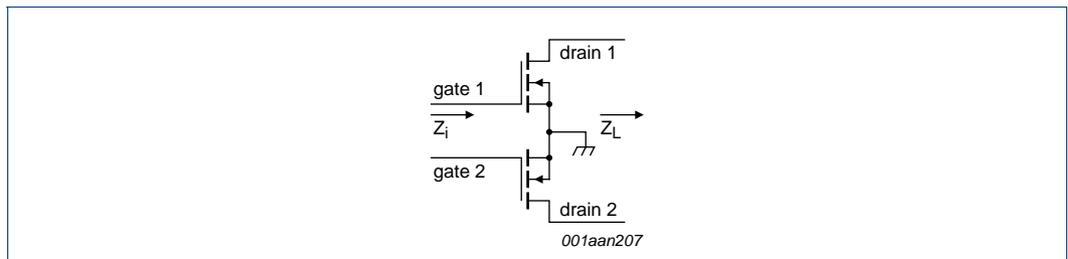
Fig 5. **Definition of transistor impedance: Z_i = input impedance, Z_L = load impedance**

3.1.2 Simulation data

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m4: 470 MHz, Zp (real) = 3.197
m2: 640 MHz, Zp (real) = 2.155
m5: 860 MHz, Zp (real) = 2.742
m8: 470 MHz, Zp (imaginary) = 0.642
m11: 640 MHz, Zp (imaginary) = −0.925
m7: 860 MHz, Zp (imaginary) = −0.343

Fig 6.   Internal drain impedance; one side, push-pull mode

019aac333

m4: 470 MHz, Zp (real) = 0.070
m2: 640 MHz, Zp (real) = 0.015
m5: 860 MHz, Zp (real) = 0.041
m8: 470 MHz, Zp (imaginary) = −7.314
m11: 640 MHz, Zp (imaginary) = −2.335
m7: 860 MHz, Zp (imaginary) = −5.641

Fig 7.   Internal drain impedance; one side, common mode

AN11062#2

Application note

Rev. 2 — 1 September 2015

8 of 20

3.2 Bias and decoupling circuit

The RF amplifier is sensitive to oscillations during mismatch. Therefore special care was taken to ensure adequate low frequency clamping at both the input and at the output. Two clamping resistors, R1 and R2, were added at the output for this purpose. Also, the supply is connected to the balun at the output. At the input, clamping is achieved via the RC network connected to the balun.

3.3 Thermal considerations

The circuit is designed to deliver 110 W DVB-T average power. The transistor can also handle CW, 2-tone and analog TV signals. However the power dissipation and heat generated by the transistor and the circuit can be excessively high, especially with CW and analog TV signals. The maximum junction temperature of the transistor is 200 °C. The electromagnetic (see Ref. 1) and lifetime (20 years minimum) restrictions must also be respected.

In this circuit, power dissipation is limited by the matching capacitors C1 and C2, and the balun (P_{max} = 150 W (average). For short-term testing, the circuit can handle CW power levels up to 450 W.

4. Test results

4.1 Large signal measurements

The test circuit was evaluated for DVB-T (110 W to 130 W average), 2-tone (250 W average) and pulsed CW. Besides the standard tests, special measurements were added with V_{DS} variation (46 V to 50 V) and water temperature variation (20 °C to 60 °C).

The demo circuit has a soldered transistor. This gives the best RF performance. A comparison is shown between a soldered and a clamped transistor.

The measurement results are shown in Figure 8 to Figure 13. All measurements were taken at V_{DD} = 50 V, I_{Dq} = 1.3 A (except for the V_{DS} variation) and T_{water} = 20 °C (except for the T_{hs}/T_{water} variation).
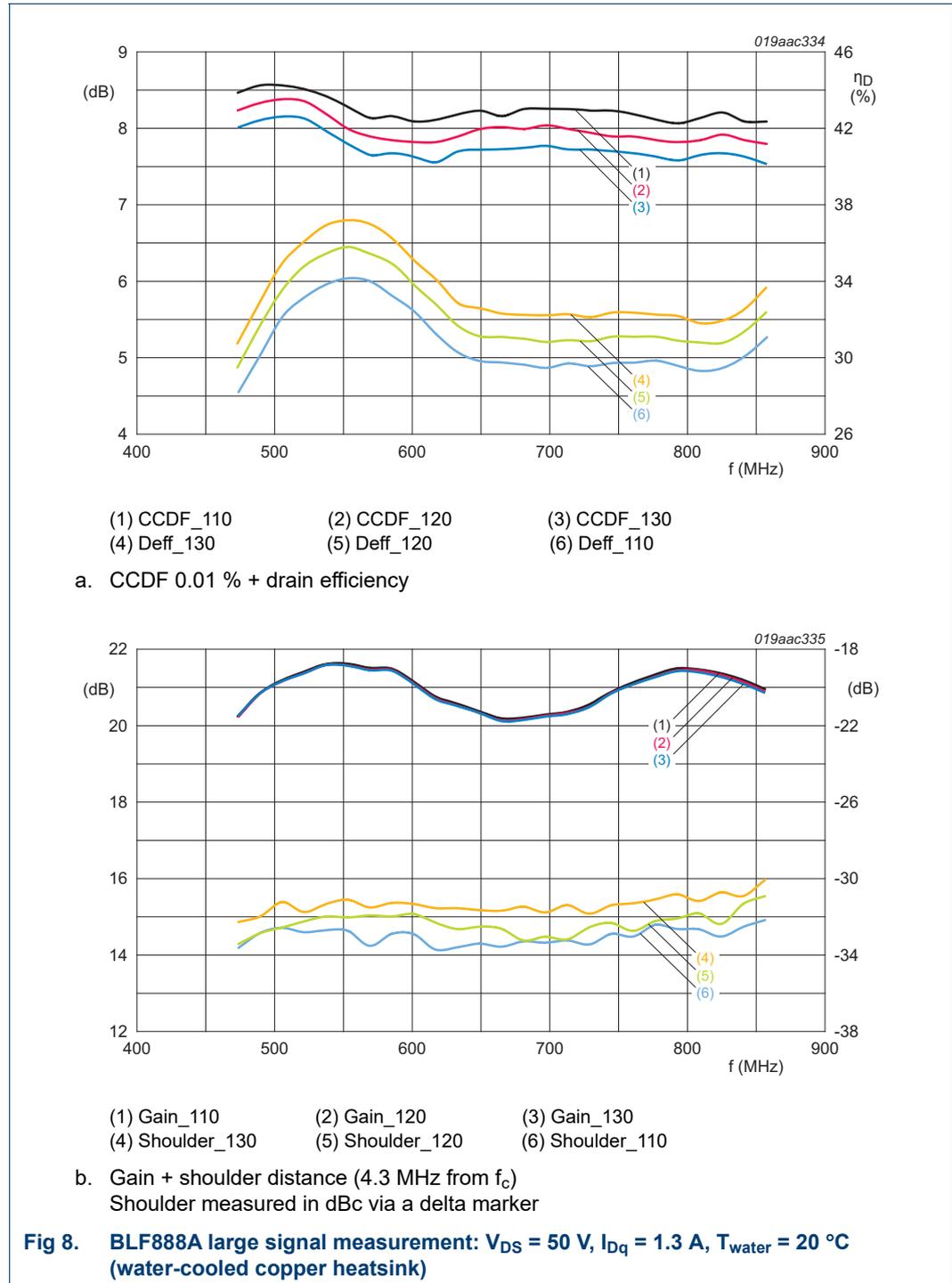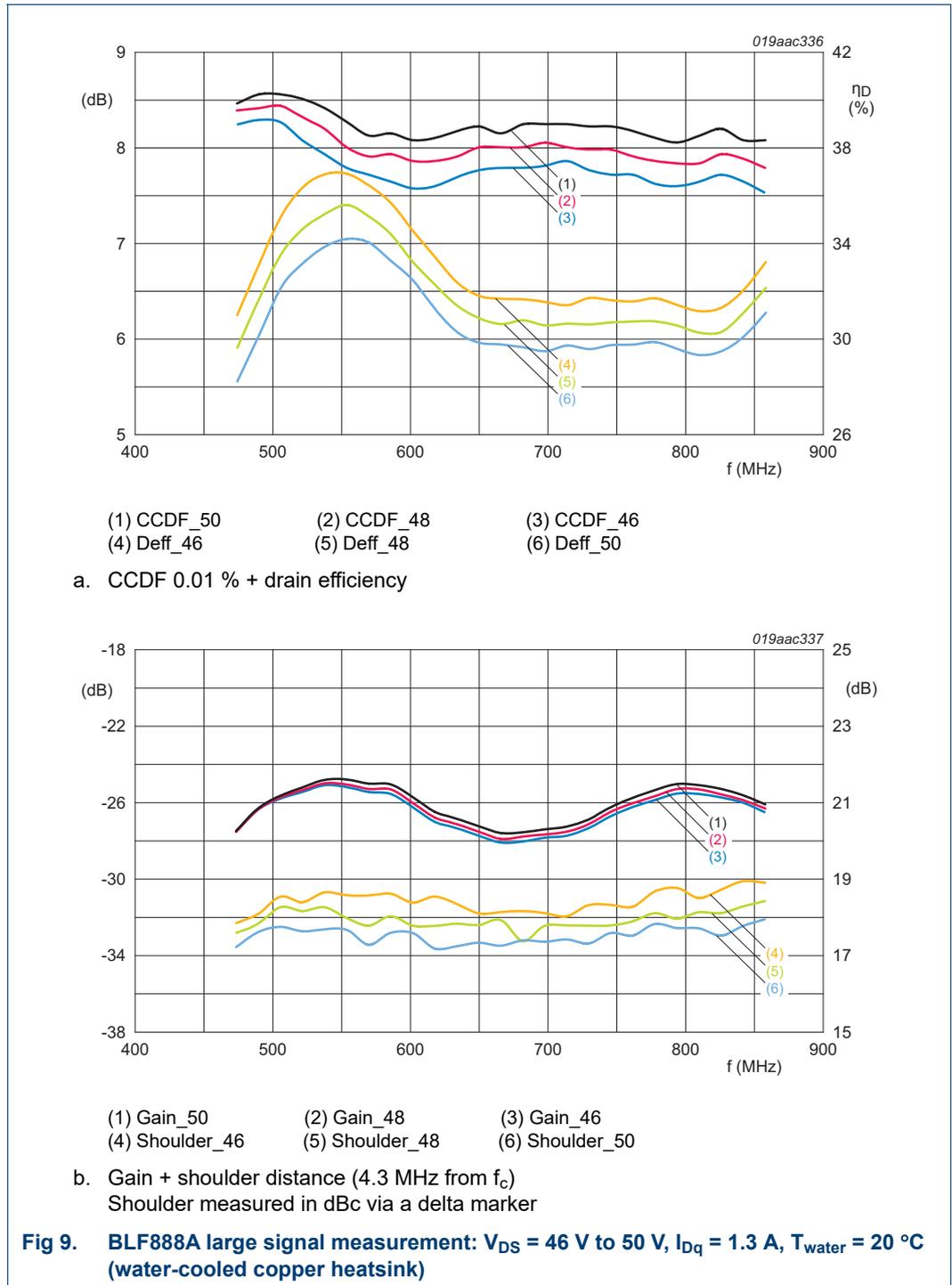
4.1.1 DVB-T

The target for the broadband circuit was to achieve average output power ($P_{L(AV)}$) of 110 W with a Peak-to-Average power Ratio (PAR) of 8 dB at a CCDF of 0.01 %. This target was met over the entire frequency range from 470 MHz to 860 MHz. Lead contact and capacitor positioning were extremely important in ensuring the target was met at the critical higher UHF frequencies (above 800 MHz). Shoulder distance is less than −30 dBc and gain is greater than 20 dB over the entire frequency band.

Drain efficiency (η_D) is greater than 28 % and is typically 30 % (at 110 W). Efficiency at the less responsive frequencies can be improved by lowering the supply voltage (the margin of the PAR at the less responsive frequencies in terms of η_D is wide enough to allow the supply voltage to be lowered).

The influence of the cooling conditions (liquid cooled) can be seen in Figure 11. The influence on CCDF / η_D is marginal in the measured range of T_{water} (20 °C to 60 °C). Gain will change by max. 1 dB.

AN11062#2

Application note **Rev. 2 — 1 September 2015** **9 of 20**

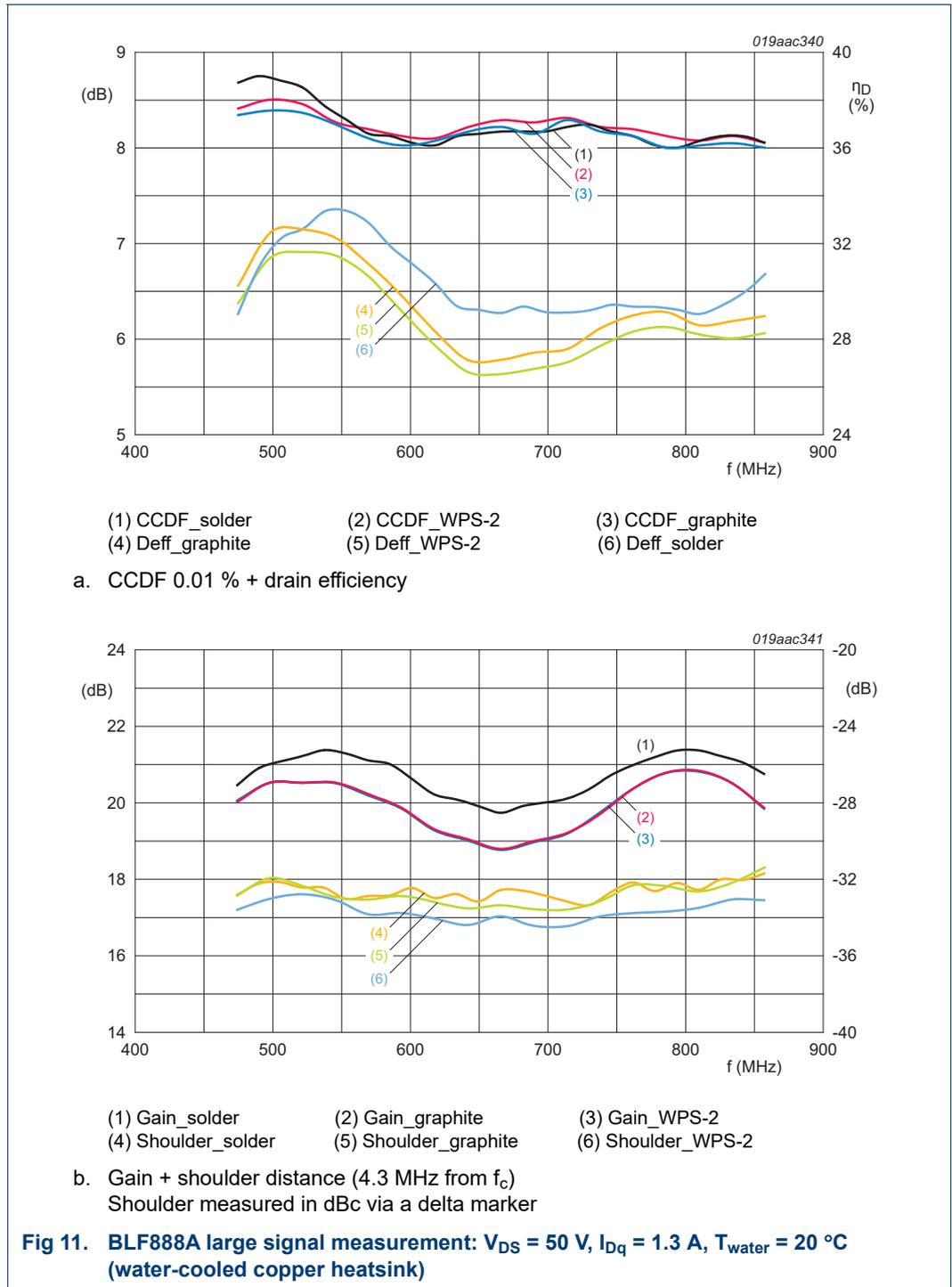Mounting has a significant effect on RF performance. This is shown in Figure 12. Obviously best figures are achieved with a soldered transistor (CCDF / η_D). The comparison (soldered / clamped) was made with 2 different compounds: graphite (Fischer thermal transfer compound WLPG) and Austerlitz WPS-2 (without silicone). The graphite compound is electrically conducting and gives slightly better thermal / RF results.

(1) CCDF_110   (2) CCDF_120   (3) CCDF_130
(4) Deff_130   (5) Deff_120   (6) Deff_110

a.   CCDF 0.01 % + drain efficiency

(1) Gain_110   (2) Gain_120   (3) Gain_130
(4) Shoulder_130   (5) Shoulder_120   (6) Shoulder_110

b.   Gain + shoulder distance (4.3 MHz from f_c)
       Shoulder measured in dBc via a delta marker

Fig 8.   **BLF888A large signal measurement: V_{DS} = 50 V, I_{Dq} = 1.3 A, T_{water} = 20 °C (water-cooled copper heatsink)**

(1) CCDF_50   (2) CCDF_48   (3) CCDF_46
(4) Deff_46   (5) Deff_48   (6) Deff_50

a.   CCDF 0.01 % + drain efficiency

(1) Gain_50   (2) Gain_48   (3) Gain_46
(4) Shoulder_46   (5) Shoulder_48   (6) Shoulder_50

b.   Gain + shoulder distance (4.3 MHz from f_c)
     Shoulder measured in dBc via a delta marker

Fig 9.   **BLF888A large signal measurement: V_{DS} = 46 V to 50 V, I_{Dq} = 1.3 A, T_{water} = 20 °C (water-cooled copper heatsink)**

AN11062#2

Application note   **Rev. 2 — 1 September 2015**   **11 of 20**

Broadband DVB-T UHF power amplifier with the BLF888A

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(dB)

η_D
(%)

(1)
(2)
(3)
(4)

(5)
(6)
(7)
(8)

f (MHz)

(1) CCDF_20 °C   (2) CCDF_40 °C   (3) CCDF_50 °C   (4) CCDF_60 °C
(5) Deff_20 °C   (6) Deff_40 °C   (7) Deff_50 °C   (8) Deff_60 °C

a.   CCDF 0.01 % + drain efficiency

019aac339

(dB)

(%)

(1)
(2)
(3)
(4)

(5)
(6)
(7)
(8)

f (MHz)

(1) CCDF_20 °C   (2) CCDF_40 °C   (3) CCDF_50 °C   (4) CCDF_60 °C
(5) Gain_20 °C   (6) Gain_40 °C   (7) Gain_50 °C   (8) Gain_60 °C

b.   Gain + CCDF 0.01 %

Fig 10.   BLF888A large signal measurement: V_{DS} = 50 V, I_{Dq} = 1.3 A, T_{water} = 20°C to 60 °C (water-cooled copper heatsink)

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(1) CCDF_solder   (2) CCDF_WPS-2   (3) CCDF_graphite
(4) Deff_graphite   (5) Deff_WPS-2   (6) Deff_solder

  a.   CCDF 0.01 % + drain efficiency

(1) Gain_solder   (2) Gain_graphite   (3) Gain_WPS-2
(4) Shoulder_solder   (5) Shoulder_graphite   (6) Shoulder_WPS-2

  b.   Gain + shoulder distance (4.3 MHz from f_c)
      Shoulder measured in dBc via a delta marker

Fig 11. **BLF888A large signal measurement: V_{DS} = 50 V, I_{Dq} = 1.3 A, T_{water} = 20 °C (water-cooled copper heatsink)**

4.1.2 2-tone

Measurements were taken at $P_{L(AV)}$ = 250 W (average). IMD3 levels are well below −30 dBc over the entire UHF frequency range. At these power levels, drain efficiency > 40 % and gain > 20 dB.

2-tone, gain and drain efficiency

(1) Efficiency
(2) Gain
(3) IMD3max

Fig 12. BLF888A large signal measurement: 2-tone broadband measurement at different $P_{L(AV)}$ levels (250 W); V_{DS} = 50 V, I_{Dq} = 1.3 A, T_{water} = 20 °C (water-cooled heatsink); Δf = 100 kHz, 2 carriers with equal amplitude
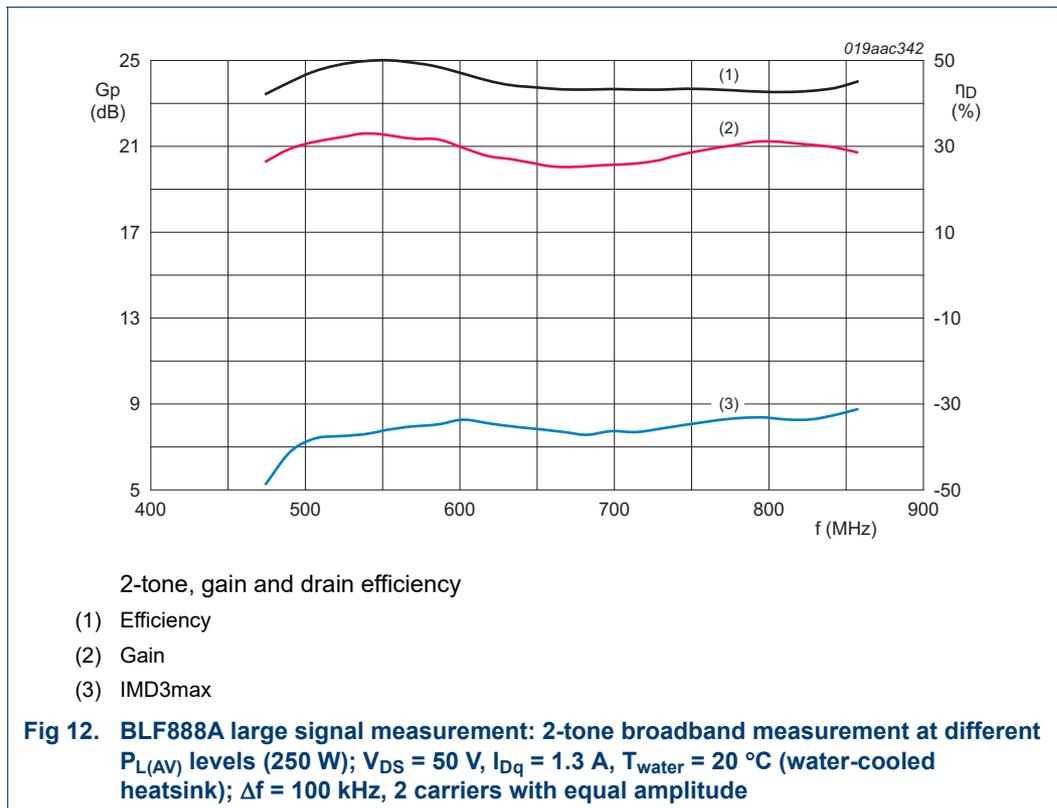
4.1.3 Pulsed

Pulsed measurement results between 470 MHz and 860 MHz (duty cycle = 10 %, t_p =100 µs) are shown in Figure 13 (response at 650 MHz and 850 MHz). This provides a clear illustration of the peak power capability of transistor plus circuit. The lowest peak power level (≈600 W) occurs at 860 MHz (due to the longer pulse duration, this is lower than the peak power with DVB-T).

AN11062#2

Application note

Rev. 2 — 1 September 2015

14 of 20

Gain + drain efficiency, t_p = 100 μs (pulse duration), δ = 10 % (duty cycle),
T = 1 ms (period)
V_{DS} = 50 V, I_{Dq} = 1.3 A, T_{water} = 20 °C (water-cooled heatsink)

(1) Power added efficiency 650 MHz
(2) Power added efficiency 860 MHz
(3) Gain 650 MHz
(4) Gain 860 MHz

Fig 13. BLF888A large signal measurement: pulsed CW broadband measurement
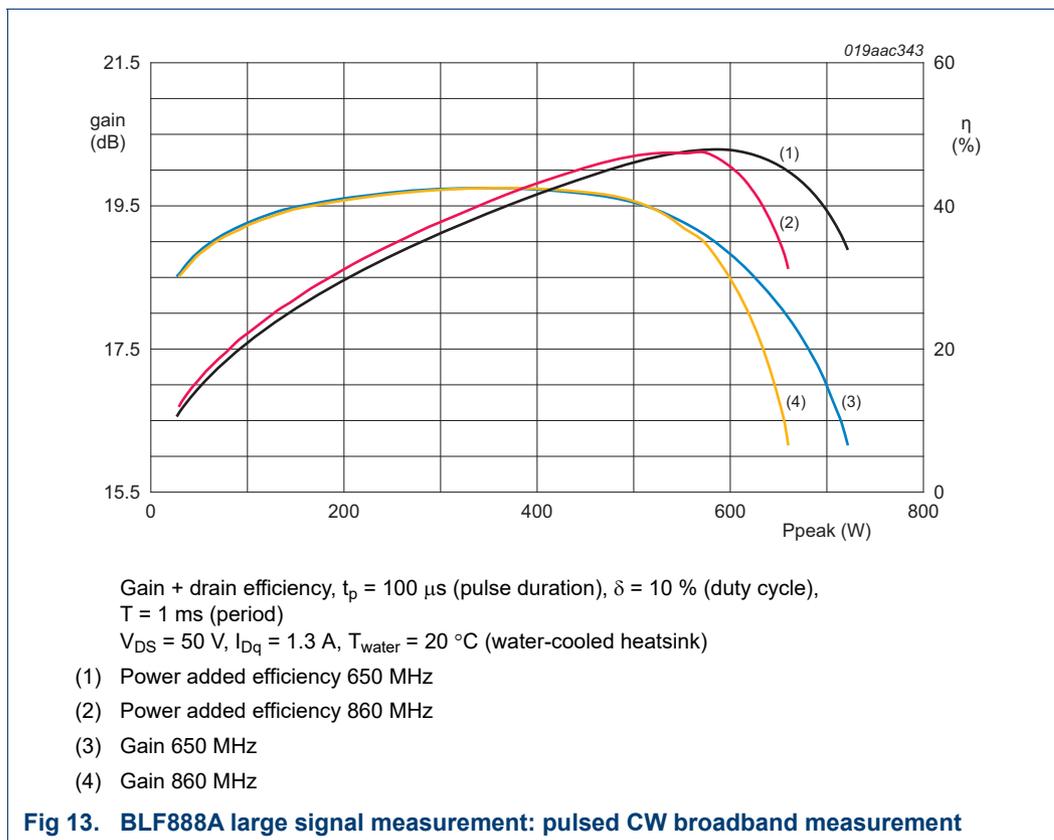
5. Conclusion

The BLF888A broadband application circuit presented in this report fulfils the following requirements:

- $P_{L(AV)}$ DVB-T > 110 W
- Efficiency > 28 % (typ. 30 %)
- Shoulder distance < −30 dBc (typ. −33 dBc)
- Gain > 20 dB (typ. 21 dB)

The circuit was designed for 110 W average DVB-T. The BLF888A is capable of delivering higher average power levels (e.g. in an ATV application). In such applications, special care must be taken to ensure adequate cooling is provided for the transistor and the circuit will need some redesign to handle the higher average power levels (e.g. the balun size).

The transistor has an extremely low R_{th} (0.15 K/W) which guarantees low junction temperatures and high reliability. It also enables the use of analog modulated TV signals.

Critical aspects of this broadband design include the harmonic loading (especially the 2nd harmonic of frequencies below 500 MHz), lead contact (an 'air gap' will shift the impedance level significantly) and low frequency stability (several damping resistors were added).

The transistor has an excellent ruggedness, with measured VSWR at 860 MHz > 40:1. Other frequencies show at least a VSWR >10:1.

6. Appendix A: PCB layout and Bill of Materials
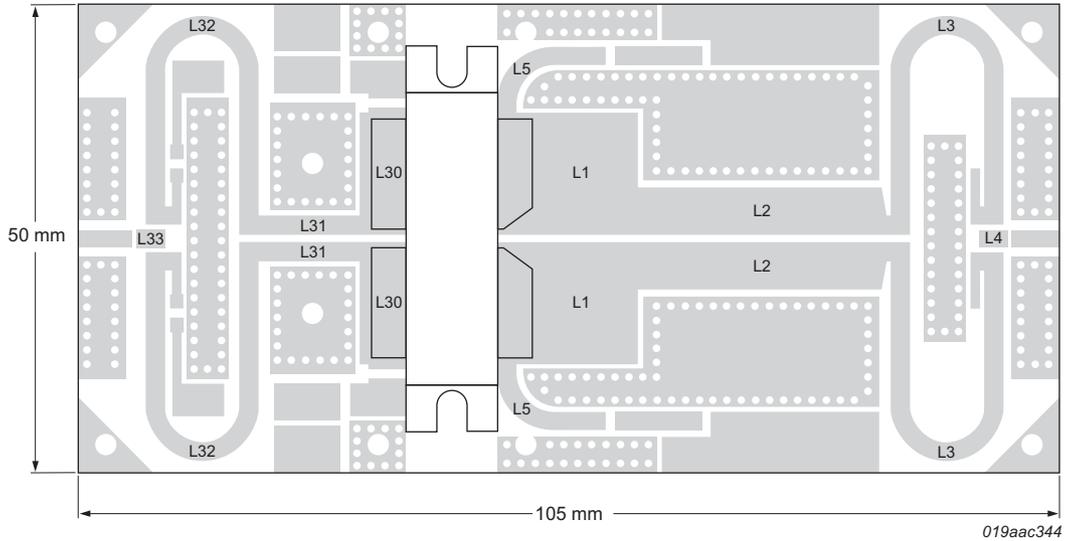
See Table 1 for a list of components.

All dimensions in mm; PCB input = 35 mm × 50 mm, PCB output = 60 mm × 50 mm, M2 holes

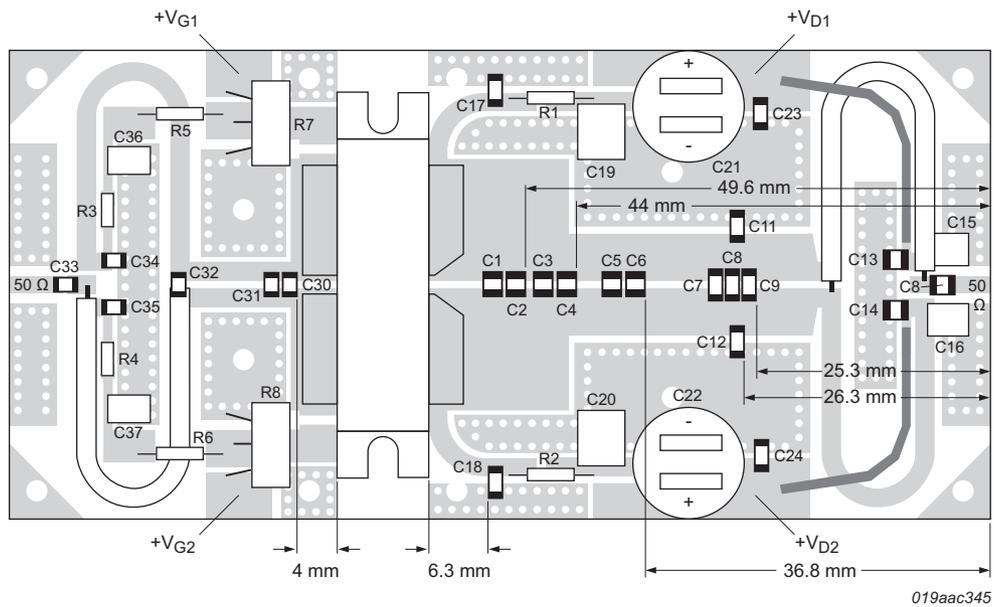**Fig 14. Dimensions of BLF888A broadband application circuit**

Fig 15. Component layout of BLF888A broadband application circuit

Table 1. **Parts list for BLF888A broadband application circuit**

Component	Value	Type	Comment
Output			
C1	12 pF	ATC180R	
C2, C3, C4, C5, C6	8.2 pF	ATC180R	
C7	6.8 pF	ATC100B	
C8	2.7 pF	ATC180R	
C9	2.2 pF	ATC100B	
C10, C13, C14	100 pF	ATC180R	
C11, C12	10 pF	ATC100B	
C17, C18, C23, C24	100 pF	ATC180B	
C15, C16	4.7 μF	TDK C4532X7R1E475MT020U	50 V
C19, C20	10 μF	TDK C570X7R1H106KT000N	50 V
C21, C22	470 μF	electrolytic capacitor	63 V
R1, R2	10 Ω		
L1	15 mm $\times$ 13 mm	microstrip line	length $\times$ width
L2	26 mm $\times$ 5 mm	microstrip line	length $\times$ width
L3	49.5 mm $\times$ 2 mm	microstrip line	length $\times$ width
L4	3.5 mm $\times$ 1.7 mm	microstrip line	length $\times$ width
L5	9.5 mm $\times$ 2 mm	microstrip line	length $\times$ width
balun B1	semi-rigid coax Z_o = 25 Ω; 49.5 mm	UT-090C-25 (EZ 90-25)	
PCB		Taconic RF35, ε_r = 3.5	
		height = 0.76 mm; 50 mm $\times$ 60 mm	
		Cu plating 35 μm	
Input			
C30	10 pF	ATC100A	
C31	9.1 pF	ATC100A	
C32	3.9 pF	ATC100A	
C33, C34, C35	100 pF	ATC100A	
C36, C37	4.7 μF	TDK C4532X7R1E475MT020U	
R3, R4	5.6 Ω		
R5, R6	100 Ω		
R7, R8	1 kΩ	potentiometer	
L30	13 mm $\times$ 5 mm	microstrip line	length $\times$ width
L31	11 mm $\times$ 2 mm	microstrip line	length $\times$ width
L32	49.5 mm $\times$ 2 mm	microstrip line	length $\times$ width
L33	3 mm $\times$ 2 mm	microstrip line	length $\times$ width
balun B2	semi-rigid coax Z_o = 25 Ω; 49.5 mm	UT-090C-25 (EZ 90-25)	

Table 1. **Parts list for BLF888A broadband application circuit** *…continued*

Component	Value	Type	Comment
PCB		Taconic RF35, ε_r = 3.5	
		h = 0.76 mm; 35 mm $\times$ 50 mm	
		Cu plating 35 μm	

7. Abbreviations

Table 2. **Abbreviations**

Acronym	Description
CCDF	Complementary Cumulative Distribution Function
DVB	Digital Video Broadcast
DVB-T	Digital Video Broadcast - Terrestrial
UHF	Ultra High Frequency
CW	Continuous Wave
IMD3	Third-Order Intermodulation Distortion
PEP	Peak Envelope Power

8. References

[1] BLF888A data sheet

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10. Contents